

Problem 7.12

$$(a) H(z) = \underbrace{(1-z^{-1})(1+z^{-2})}_{\text{MULTIPLY OUTER FACTORS}}(1+z^{-1}) \\ = (1-z^{-2})(1+z^{-2}) = 1-z^{-4}$$

$$\therefore y[n] = x[n] - x[n-4]$$

$$(b) H(e^{j\hat{\omega}}) = H(z)|_{z=e^{j\hat{\omega}}} = 1 - e^{-j4\hat{\omega}}$$

$$(c) H(e^{j\hat{\omega}}) = e^{-j2\hat{\omega}}(e^{+j2\hat{\omega}} - e^{-j2\hat{\omega}}) \\ = 2j e^{j2\hat{\omega}} \sin 2\hat{\omega} = (2\sin 2\hat{\omega}) e^{j(\pi/2 - 2\hat{\omega})}$$

MAG: $2\sin 2\hat{\omega}$

PHASE: $\frac{\pi}{2} - 2\hat{\omega}$

ALTHOUGH
THIS HAS A
SIGN CHANGE
FOR $\hat{\omega} < 0$

$$(d) \text{ BLOCK WHEN } H(e^{j\hat{\omega}}) = 0$$

$$\therefore \text{SOLVE } 2\sin 2\hat{\omega} = 0$$

$$\Rightarrow \hat{\omega} = 0, \frac{\pi}{2}, \pi, -\frac{\pi}{2}$$

(e) Need $H(e^{j\frac{\pi}{3}})$ because that is the frequency of the input.

$$H(e^{j\frac{\pi}{3}}) = \left(2\sin \frac{2\pi}{3}\right) e^{j(\frac{\pi}{2} - 2\frac{\pi}{3})} \\ = 2\left(\frac{\sqrt{3}}{2}\right) e^{j(\frac{3\pi}{6} - \frac{4\pi}{6})} \\ = -\sqrt{3} e^{-j\frac{\pi}{6}} = \sqrt{3} e^{j\pi} e^{-j\frac{\pi}{6}} = \sqrt{3} e^{j\frac{5\pi}{6}}$$

$$\therefore \text{OUTPUT IS: } y[n] = \sqrt{3} \cos\left(\frac{\pi n}{3} + \frac{5\pi}{6}\right)$$

Problem 8.3

$$y[n] = \frac{1}{2}y[n-1] + \frac{1}{3}y[n-2] - x[n] + 3x[n-1] - 2x[n-2]$$

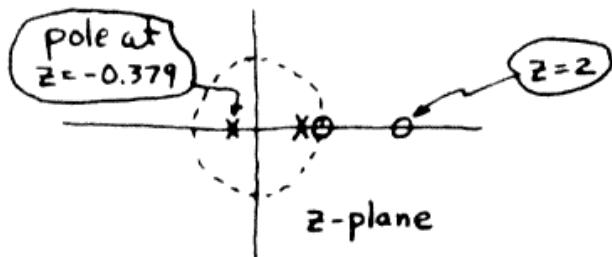
$$\overline{Y(z)} = \frac{1}{2}z^{-1}Y(z) + \frac{1}{3}z^{-2}Y(z) - \overline{X(z)} + 3z^{-1}\overline{X(z)} - 2z^{-2}\overline{X(z)}$$

$$(1 - \frac{1}{2}z^{-1} - \frac{1}{3}z^{-2})Y(z) = (-1 + 3z^{-1} - 2z^{-2})\overline{X(z)}$$

$$H(z) = \frac{Y(z)}{\overline{X(z)}} = \frac{-1 + 3z^{-1} - 2z^{-2}}{1 - \frac{1}{2}z^{-1} - \frac{1}{3}z^{-2}}$$

Change to positive powers of z to find roots.

$$H(z) = -\frac{z^2 - 3z + 2}{z^2 - \frac{1}{2}z - \frac{1}{3}} = -\frac{(z-2)(z-1)}{(z-0.879)(z+0.379)}$$



For the second system only the signs on $y[n-2]$ and $x[n-2]$ change, so we can write $H(z)$ immediately:

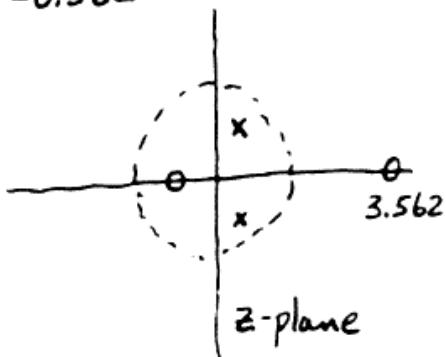
$$H(z) = -\frac{1 - 3z^{-1} - 2z^{-2}}{1 - \frac{1}{2}z^{-1} + \frac{1}{3}z^{-2}} = -\frac{z^2 - 3z - 2}{z^2 - \frac{1}{2}z + \frac{1}{3}}$$

ZEROS: $\frac{3 \pm \sqrt{9+8}}{2} = 3.562, -0.562$

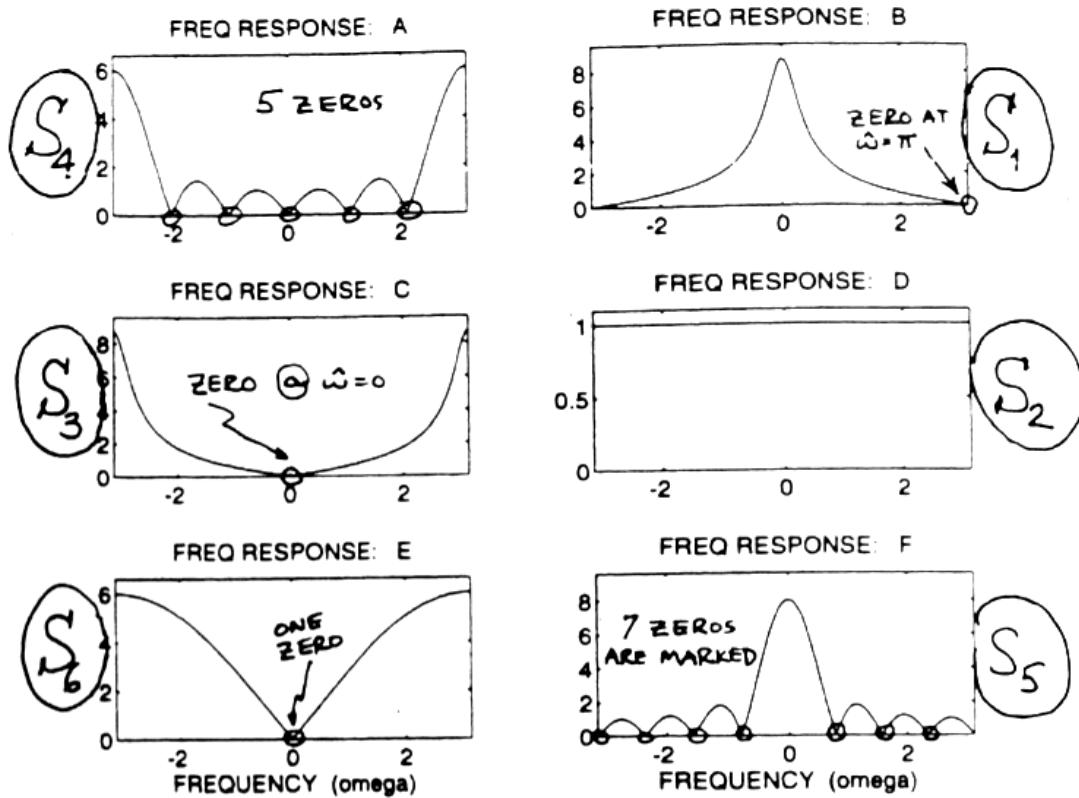
POLES: $0.25 \pm j0.52$
 $\hookrightarrow = 0.5774 e^{\pm j0.357\pi}$

ANGLE is $\pm 64.34^\circ$

or ± 1.123 rads.



Problem 8.14



For each of the frequency response plots (A, B, C, D, E, F), determine which one of the following systems (specified by either an $H(z)$ or a difference equation) matches the frequency response

$$S_1: \quad y[n] = 0.77y[n-1] + x[n] + x[n-1] \rightsquigarrow \text{LOWPASS w/ ZERO AT } \hat{\omega} = \pi$$

$$S_2: \quad y[n] = 0.77y[n-1] + 0.77x[n] - x[n-1] \rightsquigarrow H(z) = \frac{0.77 - z^{-1}}{1 - 0.77z^{-1}}, \text{ is ALL-PASS}$$

$$S_3: \quad H(z) = \frac{1 - z^{-1}}{1 + 0.77z^{-1}} \rightsquigarrow \text{HIGH-PASS w/ ZERO AT } \hat{\omega} = 0$$

$$S_4: \quad H(z) = 1 - z^{-1} + z^{-2} - z^{-3} + z^{-4} - z^{-5} \rightsquigarrow \text{HIGH PASS w/ 5 ZEROS ON U.C.}$$

$$S_5: \quad y[n] = \sum_{k=0}^7 x[n-k] \rightsquigarrow \text{LOW PASS w/ 7 ZEROS ON U.C.}$$

$$S_6: \quad H(z) = 3 - 3z^{-1} \quad \text{HIGH-PASS w/ 1 ZERO}$$

$$S_7: \quad y[n] = x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4] + x[n-5]$$

\rightsquigarrow "Dirichlet" LOWPASS w/ 5 ZEROS ON U.C.

Problem 8.18

Multiply out $H(z)$

$$\begin{aligned} H(z) &= \frac{(1-z^{-1})(1-jz^{-1})(1+jz^{-1})}{(1-0.9e^{j2\pi/3}z^{-1})(1-0.9e^{-j2\pi/3}z^{-1})} \\ &= \frac{(1-z^{-1})(1+z^{-2})}{1-2(0.9)\cos(\frac{2\pi}{3})z^{-1}+(0.9)^2z^{-2}} \\ &= \frac{1-z^{-1}+z^{-2}-z^{-3}}{1-0.9z^{-1}+0.81z^{-2}} \end{aligned}$$

(a) Use the numerator & denominator polynomial coefficients as filter coefficients:

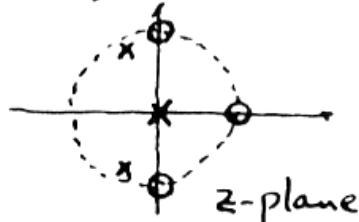
$$y[n] = 0.9y[n-1] - 0.81y[n-2] + x[n] - x[n-1] + x[n-2] - x[n-3]$$

(b) Multiply numerator & denominator by z^3 :

$$H(z) = \frac{(z-1)(z-j)(z+j)}{z(z-0.9e^{j2\pi/3})(z-0.9e^{-j2\pi/3})}$$

Zeroes: $z=1, j$ and $-j$

Poles: $z=0, z=0.9e^{\pm j2\pi/3}$



(c) The zeros of the numerator polynomial are on the unit circle at $z=e^{j0}$, $z=e^{j\pi/2}$ and $z=e^{-j\pi/2}$

When $x[n] = Ae^{j\varphi}e^{j\omega n}$, the output $y[n]$ is

$$y[n] = H(e^{j\hat{\omega}}) \cdot Ae^{j\varphi} e^{j\hat{\omega}n}$$

There the output will be zero when $H(e^{j\hat{\omega}})=0$.

That is, for $\hat{\omega}=0$, $\hat{\omega}=\pi/2$ and $\hat{\omega}=-\pi/2$.